

# Appendix E

## The Signal Term for PeiDLL

In this appendix, an expression for the signal term  $s_m^{pel}$  for the PeiDLL scheme, which is defined in (3.18), is obtained. This is derived as follows.

$$\begin{aligned}
 s_m^{pel} &= \frac{1}{T_c} \int_{(q-n)T_c}^{qT_c} c(t-\tau) c_m^{pel}(t) dt \\
 &= \frac{1}{T_c} \int_{(q-n)T_c}^{qT_c} c(t-\beta T_c - \gamma T_c) [c(t-2mT_c - \delta T_c) - c(t-2mT_c + \delta T_c)] dt \\
 &= s_m^{pel,+} - s_m^{pel,-}
 \end{aligned} \tag{E.1}$$

where  $s_m^{pel,+}$  is the chip normalized partial correlation between the received PN sequence and the delayed version of local PN sequence in the  $m^{\text{th}}$  branch given by

$$s_m^{pel,+} = \frac{1}{T_c} \int_{(q-n)T_c}^{qT_c} c(t-\beta T_c - \gamma T_c) c(t-2mT_c - \delta T_c) dt, \tag{E.2}$$

and  $s_m^{pel,-}$  is the chip normalized partial correlation between the received PN sequence and the advanced version of local PN sequence in the  $m^{\text{th}}$  branch, given by

$$s_m^{pel,-} = \frac{1}{T_c} \int_{(q-n)T_c}^{qT_c} c(t-\beta T_c - \gamma T_c) c(t-2mT_c + \delta T_c) dt. \tag{E.3}$$

The signals  $s_m^{pel,+}$  and  $s_m^{pel,-}$  can be computed as follows.

**E.1) Computing the signal  $s_m^{pel,+}$ .** This term can be simplified as

$$\begin{aligned}
 s_m^{pel,+} &= \frac{1}{T_c} \sum_{i=q-n}^{q-1} \int_{iT_c}^{(i+1)T_c} \sum_{j=-\infty}^{\infty} c_j P_{T_c}(t-\beta T_c - \gamma T_c - jT_c) \sum_{k=-\infty}^{\infty} c_k P_{T_c}(t-2mT_c - \delta T_c - kT_c) dt \\
 &= \frac{1}{T_c} \sum_{i=q-n}^{q-1} s_{m,i}^{pel,+}
 \end{aligned} \tag{E.4}$$

where  $s_{m,i}^{pel,+}$  is the correlation over one chips duration. Figure E.1 shows the timing relations for the local and received PN code, with  $\delta > \gamma$  and  $\delta < \gamma$ . As shown in the figure, the integration during one period can be broken down into three terms.

E.1.1) For  $\delta > \gamma$ . The signal  $s_{m,i}^{pel,+}$  can be written as

$$s_{m,i}^{pel,+} = \gamma T_c c_{i-2m-1} c_{i-\beta-1} + (\delta - \gamma) T_c c_{i-2m-1} c_{i-\beta} + (1 - \delta) T_c c_{i-2m} c_{i-\beta}. \tag{E.5}$$

E.1.2) For  $\delta < \gamma$ . The signal  $s_{m,i}^{pel,+}$  can be written as

$$s_{m,i}^{pel,+} = \delta T_c c_{i-2m-1} c_{i-\beta-1} + (\gamma - \delta) T_c c_{i-2m} c_{i-\beta-1} + (1 - \gamma) T_c c_{i-2m} c_{i-\beta}. \tag{E.6}$$

With some mathematical manipulation, we can write (E.5) and (E.6) as one expression:

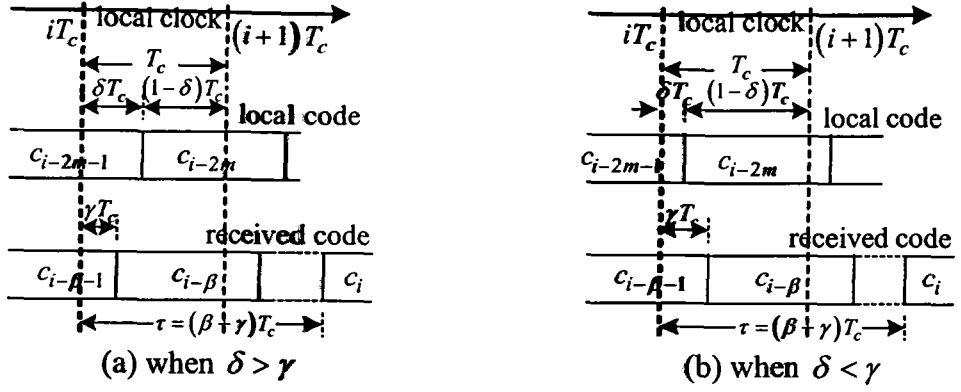


Figure E.1. The relation between the delayed version of local code and the received signal at the  $m^{\text{th}}$  branch during  $iT_c \leq t \leq (i+1)T_c$ .

$$\begin{aligned}
 S_{m,i}^{pel,+} = & \left( \frac{\delta + \gamma}{2} - \frac{|\delta - \gamma|}{2} \right) T_c c_{i-2m-1} c_{i-\beta-1} + \left( \frac{\delta - \gamma}{2} + \frac{|\delta - \gamma|}{2} \right) T_c c_{i-2m-1} c_{i-\beta} \\
 & + \left( \frac{\gamma - \delta}{2} + \frac{|\gamma - \delta|}{2} \right) T_c c_{i-2m} c_{i-\beta-1} + \left( \frac{2 - \gamma - \delta}{2} - \frac{|\delta - \gamma|}{2} \right) T_c c_{i-2m} c_{i-\beta}. \quad (E.7)
 \end{aligned}$$

**E.2) Computing the signal  $s_{m,i}^{pel,-}$ .** This term can be written as

$$\begin{aligned}
 S_m^{pel,-} &= \frac{1}{T_c} \sum_{i=q-n}^{q-1} \int_{iT_c}^{(i+1)T_c} \sum_{j=-\infty}^{\infty} c_j P_{T_c}(t - \beta T_c - \gamma T_c - j T_c) \sum_{k=-\infty}^{\infty} c_k P_{T_c}(t - 2m T_c + \delta T_c - k T_c) dt \\
 &= \frac{1}{T_c} \sum_{i=q-n}^{q-1} S_{m,i}^{pel,-} \quad (E.8)
 \end{aligned}$$

where  $s_{m,i}^{pel,-}$  is the correlation over one chip duration. Figure E.2 shows the timing for the local and received PN codes with  $\delta > \gamma$  and  $\delta < \gamma$ . Similarly, the integration can be broken into three subsections as follows.

E.2.1) For  $\delta < \gamma$  or  $(1-\delta) > \gamma$ . The signal  $s_{m,i}^{pel,-}$  can be written as

$$S_{m,i}^{pel,-} = \gamma T_c c_{i-2m} c_{i-\beta-1} + (1 - \delta - \gamma) T_c c_{i-2m} c_{i-\beta} + \delta T_c c_{i-2m+1} c_{i-\beta}. \quad (E.9)$$

E.2.2) For  $\delta > \gamma$  or  $(1-\delta) < \gamma$ . The signal  $s_{m,i}^{pel,-}$  can be written as

$$S_{m,i}^{pel,-} = (1 - \delta) T_c c_{i-2m} c_{i-\beta-1} + (\gamma + \delta - 1) T_c c_{i-2m-1} c_{i-\beta-1} + (1 - \gamma) T_c c_{i-2m+1} c_{i-\beta}. \quad (E.10)$$

Using some mathematical manipulation, a unified expression can be obtained:

$$\begin{aligned}
 S_{m,i}^{pel,-} = & \left( \frac{(1-\delta) + \gamma}{2} - \frac{|(1-\delta) - \gamma|}{2} \right) T_c c_{i-2m} c_{i-\beta-1} + \left( \frac{(1-\delta) - \gamma}{2} + \frac{|(1-\delta) - \gamma|}{2} \right) T_c c_{i-2m} c_{i-\beta} \\
 & + \left( \frac{\gamma - (1-\delta)}{2} + \frac{|\gamma - (1-\delta)|}{2} \right) T_c c_{i-2m+1} c_{i-\beta-1} + \left( \frac{1 + \delta - \gamma}{2} - \frac{|(1-\delta) - \gamma|}{2} \right) T_c c_{i-2m+1} c_{i-\beta}. \quad (E.11)
 \end{aligned}$$

Finally, from (E.1), (E.4), (E.7), (E.8), and (E.11), a complete expression for the signal  $s_m^{pel}$  can be written as

$$S_m^{pel} = \sum_{i=0}^{n-1} \left[ \left( \frac{\delta + \gamma}{2} - \frac{|\delta - \gamma|}{2} \right) c_{q-n+i-2m-1} c_{q-n+i-\beta-1} + \left( \frac{\delta - \gamma}{2} + \frac{|\delta - \gamma|}{2} \right) c_{q-n+i-2m-1} c_{q-n+i-\beta} \right]$$

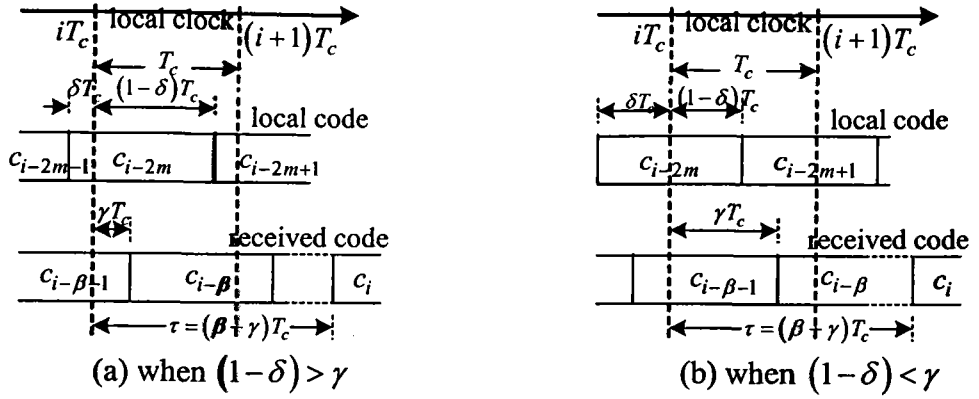


Figure E.2. The relation between the advanced version of local code and the received signal at the  $m^{\text{th}}$  branch during  $iT_c \leq t \leq (i+1)T_c$ .

$$\begin{aligned}
 & + \left( \frac{|(1-\delta)-\gamma|}{2} + \frac{|\gamma-\delta|}{2} - \frac{1}{2} \right) c_{q-n+i-2m} c_{q-n+i-\beta-1} + \left( \frac{1}{2} - \frac{|\delta-\gamma|}{2} - \frac{|(1-\delta)-\gamma|}{2} \right) c_{q-n+i-2m} c_{q-n+i-\beta} \\
 & - \left( \frac{\gamma-(1-\delta)}{2} + \frac{|\gamma-(1-\delta)|}{2} \right) c_{q-n+i-2m+1} c_{q-n+i-\beta-1} \\
 & - \left. \left( \frac{1+\delta-\gamma}{2} - \frac{|(1-\delta)-\gamma|}{2} \right) c_{q-n+i-2m+1} c_{q-n+i-\beta} \right]. \tag{E.12}
 \end{aligned}$$