

Appendix J

Variance of $w_{m,L}$

In this appendix, we derive the variance $\sigma_{m,L}^2$ of $w_{m,L}$, for $m \in \{1, 2, 3, \dots, 2^M\}$. It is defined as

$$\sigma_{m,L}^2 = E\{w_{m,L}w_{m,L}\} - (E\{w_{m,L}\})^2 \quad (J.1)$$

Substituting $E\{w_{m,L}\} = \sqrt{2PL}T_c R_m(\tau)$ and (4.15) into (J.1) yields

$$\begin{aligned} \sigma_{m,L}^2 &= E\left\{\left(\sqrt{2PL}T_c R_m(\tau)\right)\left(\sqrt{2PL}T_c R_m(\tau)\right)\right\} + 2E\left\{\sqrt{2PL}T_c R_m(\tau)\eta_{m,L}\right\} \\ &\quad + E\{\eta_{m,L}\eta_{m,L}\} - \left(\sqrt{2PL}T_c R_m(\tau)\right)\left(\sqrt{2PL}T_c R_m(\tau)\right) \\ &= 2E\left\{\sqrt{2PL}T_c R_m(\tau)\eta_{m,L}\right\} + E\{\eta_{m,L}\eta_{m,L}\}. \end{aligned} \quad (J.2)$$

Since $E\{\eta_{m,L}\} = 0$, $\sigma_{m,L}^2$ is equal to $E\{\eta_{m,L}\eta_{m,L}\}$, which can be computed as

$$\begin{aligned} E\{\eta_{m,L}\eta_{m,L}\} &= E\left\{\left[\sum_{i=0}^{L-1} \int_{iT_c}^{(i+1)NT_c} \text{Re}\{\eta(t)\} a_m(t) dt\right] \left[\sum_{j=0}^{L-1} \int_{jNT_c}^{(j+1)NT_c} \text{Re}\{\eta(\lambda)\} a_m(\lambda) d\lambda\right]\right\} \\ &= \sum_{i=0}^{L-1} \sum_{j=0}^{L-1} \int_{iT_c}^{(i+1)NT_c} \int_{jNT_c}^{(j+1)NT_c} E\{\text{Re}\{\eta(t)\} \text{Re}\{\eta(\lambda)\}\} a_m(t) a_m(\lambda) d\lambda dt. \end{aligned} \quad (J.3)$$

The autocorrelation function of the real part of noise is

$$E\{\text{Re}\{\eta(t)\} \text{Re}\{\eta(\lambda)\}\} = N_0 \delta(t - \lambda) \quad (J.4)$$

where $\delta(t)$ is a delta function. Substituting (J.4) into (J.3) and recalling that

$a_m(t) = \sum_{i=0}^{N-1} a_{m,i} P_{T_c}(t - iT_c)$, we obtain

$$\begin{aligned} E\{\eta_{m,L}\eta_{m,L}\} &= N_0 \sum_{i=0}^{L-1} \int_{iT_c}^{(i+1)NT_c} a_m(t) a_m(t) dt \\ &= N_0 L T_c \sum_{i=0}^{N-1} a_{m,i} a_{m,i}. \end{aligned} \quad (J.5)$$

Therefore, variance of $\eta_{m,L}$ can be expressed as

$$\sigma_{m,L}^2 = N_0 L T_c \sum_{i=0}^{N-1} a_{m,i} a_{m,i}. \quad (J.6)$$

From Figure 4.3, the autocorrelation of $a_{m,i}$, $AC(m) = \sum_{i=0}^{N-1} a_{m,i} a_{m,i}$, is a summation of two terms given below. Recall that $a_{m,i}$ comprises $(N-1)/2$ PN signals of different

code phases with positively weighted coefficients and $(N+1)/2$ different PN signals of different code phases with negatively weighted coefficients.

(i) Correlation between $a_{m,i}$ and $(N-1)/2$ PN signals of different code phases with positively weighted coefficients. Let CC_+ denote this correlation. It can be computed as

$$CC_+ = (N) \left(\frac{N-1}{2} \right) + (-1) \left[\left(\frac{N-1}{2} \right) \left(\frac{N-1}{2} \right) - \left(\frac{N-1}{2} \right) \right] + (-1)(-1) \left(\frac{N+1}{2} \right) \left(\frac{N-1}{2} \right) \quad (J.7)$$

$$= \frac{2N^2 + 2N - 4}{4}. \quad (J.8)$$

The first and second terms of (J.7) are obtained from the correlations between $(N-1)/2$ positively weighted code phases of the PN signal due to $a_{m,i}$ and $(N-1)/2$ positively weighted code phases of the PN signal. Among them, there are $(N-1)/2$ terms with in-phase autocorrelation, contributing to the first term. The remainings, $\left[\left(\frac{N-1}{2} \right) \left(\frac{N-1}{2} \right) - \left(\frac{N-1}{2} \right) \right]$ terms produce out-of-phase autocorrelation values, contributing to the second term. The third term is obtained from the correlations between $(N+1)/2$ negatively weighted code phases of the PN signal due to $a_{m,i}$ and $(N-1)/2$ positively weighted code phases of the PN signal. These $\left(\frac{N+1}{2} \right) \left(\frac{N-1}{2} \right)$ terms give out-of-phase autocorrelation values. The negative sign (-1) in the third term is due to the product of negative and positive coefficients.

(ii) Correlation between $a_{m,i}$ and $(N+1)/2$ PN signals of different code phases with negatively weighted coefficients. Let CC_- denote the result. It can be computed as

$$CC_- = (N) \left(\frac{N+1}{2} \right) + (-1) \left[\left(\frac{N+1}{2} \right) \left(\frac{N+1}{2} \right) - \left(\frac{N+1}{2} \right) \right] + (-1)(-1) \left(\frac{N-1}{2} \right) \left(\frac{N+1}{2} \right) \quad (J.9)$$

$$= \frac{2N^2 + 2N}{4}. \quad (J.10)$$

The terms in (J.9) are obtained similarly to (J.7). The first and second terms are from the correlations between $(N+1)/2$ negatively weighted code phases of the PN signal due to $a_{m,i}$ and $(N+1)/2$ negatively weighted code phases of the PN signal. The third term is from the correlations between $(N-1)/2$ positively weighted code phases of the PN signal due to $a_{m,i}$ and $(N+1)/2$ negatively weighted code phases of the PN signal.

Since $AC(m) = CC_+ + CC_-$, substituting (J.8) and (J.10) into (J.6) yields

$$\sigma_{m,L}^2 = N_0 L T_c \left[\left(\frac{2N^2 + 2N - 4}{4} \right) + \left(\frac{2N^2 + 2N}{4} \right) \right]$$

$$= N_0 L T_c (N^2 + N - 1). \quad (J.11)$$